

Quasi-interpolation by means of filter-banks

Gerardo Pérez-Villalón

Abstract—We consider the problem of approximating a regular function $f(t)$ from its samples, $f(nT)$, taken in a uniform grid. Quasi-interpolation schemes approximate $f(t)$ with a dilated version of a linear combination of shifted versions of a kernel $\varphi(t)$, specifically $f_{\text{approx}}^T(t) = \sum a_f[n]\varphi(t/T - n)$, in a way that the polynomials of degree at most $L - 1$ are recovered exactly. These approximation schemes give order L , i.e., the error is $O(T^L)$ where T is the sampling period. Recently, quasi-interpolation schemes using a discrete prefiltering of the samples $f(nT)$ to obtain the coefficients $a_f[n]$, have been proposed. They provide tight approximation with a low computational cost. In this work, we generalize considering rational filter banks to prefilter the samples, instead of a simple filter. This generalization provides a greater flexibility in the design of the approximation scheme. The upsampling and downsampling ratio r of the rational filter bank, plays a significant role. When $r = 1$ the scheme has similar characteristics to those related to a simple filter. Approximation schemes corresponding to smaller ratios, give less approximation quality but, in return, they have less computational cost and involve less storage load in the system.

Index Terms—Quasi-interpolation, filter bank, shift-invariant spaces, approximation order, Strang-Fix conditions.

I. INTRODUCTION

This work deals with the problem of approximating a regular function $f(t)$ from its samples taken in a uniform grid $\{f(nT)\}_{n \in \mathbb{Z}}$. In many approximation schemes the function $f(t)$ is approximated by

$$f_{\text{approx}}^T(t) = \sum_{n \in \mathbb{Z}} a[n]\varphi\left(\frac{t}{T} - n\right) \quad (1)$$

where φ is a suitable kernel and the coefficients $a[n]$ are obtained by a discrete filtering of the sequence of samples $s_f[n] = f(nT)$

$$a[n] = (s_f * g)[n], \quad (2)$$

with an advisable filter g . Equivalently,

$$f_{\text{approx}}^T(t) = \sum_{n \in \mathbb{Z}} f(nT)S\left(\frac{t}{T} - n\right) \quad (3)$$

where $S(t) = \sum_{n \in \mathbb{Z}} g[n]\varphi(t - n)$ is the reconstruction function.

For interpolation schemes (see [1], [2], [3], [4], [5]), the interpolation condition

$$f_{\text{approx}}^T(nT) = f(nT), \quad n \in \mathbb{Z}, \quad (4)$$

This work has been supported by the grant MTM2009–08345 from the D.G.I. of the Spanish Ministerio de Ciencia e Innovación (MICINN).

Copyright (c) 2008 IEEE. Personal use of this material is permitted. However, permission to use this material for any other purpose must be obtained from the IEEE by sending a request to pubs-permissions@ieee.org.

G. Pérez-Villalón is with the Departamento de Matemática Aplicada a la Telecomunicación, E.U.I.T.T, Universidad Politécnica de Madrid, 28031 Madrid, Spain (e-mail: gperez@uitt.upm.es)

is required, and then, the prefilter is necessarily given by $G(z) = (\sum \varphi(n)z^{-n})^{-1}$, where $G(z)$ denotes the z-transform of $g[n]$, $G(z) = \sum_{n \in \mathbb{Z}} g[n]z^{-n}$. The classical example is the Shannon sampling formula, where $\varphi(t) = S(t) = \text{sinc}(t)$, which exactly recovers a bandlimited function $f(t)$ taking a small enough sampling period. This formula gives a tight approximation for functions which are approximately bandlimited. The slow decay of $\text{sinc}(t)$ is its main inconvenience. A classical interpolating alternative is to consider a kernel φ of compact support, satisfying the Strang-Fix conditions of order L , i.e.,

$$\widehat{\varphi}(0) \neq 0, \widehat{\varphi}^{(l)}(n) = 0, l = 0, 1, \dots, L-1, n \in \mathbb{Z} \setminus \{0\}, \quad (5)$$

where $\widehat{\varphi}(w) = \int \varphi(t)e^{-2\pi i w t} dt$ denotes the Fourier transform. These conditions imply, under appropriate hypotheses, that the interpolation formula (3) exactly recovers any polynomial in Π_{L-1} , the space of polynomials of degree less than or equal to $L - 1$. A suitable choice for the kernel φ , e.g., a B-spline (see [6]), allows us to efficiently compute the approximation by using (2) and (1), since $g[n]$ decays exponentially fast and φ has compact support. Besides, the ability of reproducing polynomials in Π_{L-1} guarantees that the interpolation scheme has approximation order L ; specifically, for any f in the Sobolev space $W_2^L(\mathbb{R}) := \{f : \|f^{(l)}\|_2 < \infty, l \leq L\}$ (see [2], [7], [8]),

$$\|f - f_{\text{approx}}^T\|_2 = O(T^L) \quad (6)$$

and for any f in $W_\infty^L(\mathbb{R}) := \{f : \|f^{(l)}\|_\infty < \infty, l \leq L\}$ (see [8], [9]),

$$\|f - f_{\text{approx}}^T\|_\infty = O(T^L).$$

Quasi-interpolation schemes (see [7], [8], [10], [11]) of order L require only that the formula (3) reproduces polynomials in Π_{L-1} (quasi-interpolation condition), which allows us to maintain the approximation order (6). This is a weaker condition than the interpolation condition (4), giving more flexibility for the choice of the filter g . For instance, if we look for a FIR filter $G(z)$ the only condition that has to be satisfied is that the non null coefficients of the filter $G(z)$ satisfy a certain linear system of L equations (see [7], [11]). Thus, in general, we can design FIR filters g with a support of size L satisfying the aforementioned condition. It is important to note the possibility of designing quasi-interpolation schemes, with an approximation quality close to that given by the orthogonal projector onto $V_T(\varphi) = \{\sum a[n]\varphi(t/T - n) : a \in \ell^2\}$ (see [7], [12], [13], [14]).

The interpolation and quasi-interpolation schemes, have relevant applications in image processing, such as rescaling, volume rendering, evaluating gradients, translations, rotations, and in general, geometric transformations of the signal (see [4] and references therein).

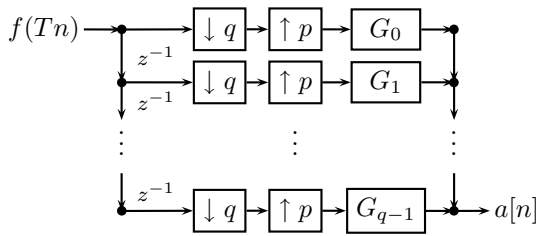


Fig. 1. Block diagram of the proposed prefiltering of the samples

In this work we study a generalization of the aforementioned quasi-interpolation schemes. We consider the possibility of obtaining quasi-interpolation schemes of order L by using a rational filter bank instead of a simple filter. Specifically, we select two natural numbers, q and p , and q discrete filters, g_0, g_1, \dots, g_{q-1} . The samples are filtered with a filter bank as shown in figure 1, yielding to a coefficient sequence $a[n]$ and then, the approximation is obtained by using

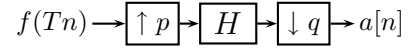
$$f_{\text{approx}}^T(t) = \sum_{n \in \mathbb{Z}} a[n] \varphi\left(\frac{p}{qT}t - n\right) \quad (7)$$

When p and q are coprime, the prefiltering can be easier represented by using the filter $H(z) = \sum_{j=0}^{q-1} z^{-jp} G_j(z^q)$ as shown in figure 2. Reciprocally, any prefiltering of the type given in figure 2 can be represented by using the filters $g_j[n] = h[nq + jp]$, $j = 0, \dots, q-1$, as shown in figure 1 (see Section II). The representation of the prefiltering with $H(z)$ is simpler than the representation with the filters $G_j(z)$. However, the reconstruction functions, the quasi-interpolation condition and the error for this approximation scheme, except in the case $p = 1$, admit a simpler expression in terms of the filters $G_j(z)$.

The proposed prefiltering implies a rational rate change, where the upsampling and downsampling ratio is $r = p/q$. Thus, we have the possibility of choosing the ratio r according to the desirable characteristics of the approximation scheme. If $r = 1$ the scheme has similar characteristics to those related to a simple filter, in fact when $p = q = 1$ it is one of them. If $r < 1$, the approximation scheme, as a consequence of the compression, loses in quality approximation (see Section IV), but it gains in numerical efficiency, since the computational cost of the algorithm is approximately proportional to r (see Sections II and III). Note that the compression associated with a ratio $r < 1$ is also suitable in order to transmit or store the signal. From the practical point of view the case of $r > 1$, although theoretically possible, is not interesting.

In [15] one can find conditions on the filters G_j (with a slight change of notation) that guarantee the perfect reconstruction of the functions belonging to $V_T(\varphi)$. This perfect reconstruction is not required here; it is only required for polynomials in Π_{L-1} (a weaker condition). Thus, we gain in flexibility, allowing us to design approximation schemes involving filters with a smaller support.

In [16] a variational method which applies to this setting is given. It allows to approximate a function from non-uniform samples. For $r < 1$ and without variational penalty, the method


 Fig. 2. Prefiltering when p and q are coprime

gives the function in $V_T(\varphi)$ which is the closest to the samples in the ℓ^2 sense.

For the sake of simplicity, in this work, we confine ourselves to FIR filters. However, most of the results are valid to IIR filters (with suitable decay conditions). For the case of $r = 1$, in [7], [12], [14] are given IIR filter satisfying the quasi-interpolating condition, which can be implemented by a fast algorithm. See [13] for the case of $r = 1/q$.

The paper is organized as follows: In Section II we introduce the proposed approximation scheme. In Section III, we deduce a necessary and sufficient condition (see infra (16)) on this scheme to reproduce polynomials of degree at most $L - 1$ (quasi-interpolation condition), and we show that by solving a linear system of equations it is possible to obtain FIR filters G_j satisfying this condition. In Section IV we give examples of such FIR filters. In Section V we prove that, whenever the aforementioned condition holds, the scheme has approximation order L in both, the $L^2(\mathbb{R})$ and $L^\infty(\mathbb{R})$ -norm sense. In this section, by applying the error formula given in [10], we get an expression for the approximation error, which gives us a suitable criterium in order to choose among the filters G_j that satisfy the quasi-interpolation condition. In Section VI we give simulations showing the behavior of the approximation technique for different values of the ratio r . We conclude in Section VII. An appendix includes the proofs of the results given in Sections II, III and V.

II. THE APPROXIMATION SCHEME

Throughout this paper we assume that φ is a continuous real function, compactly supported, that satisfies the Strang-Fix conditions of order L (5). We also assume that the sequence $\{\varphi(t - n)\}_{n \in \mathbb{Z}}$ is a Riesz basis for the space $V_1(\varphi) = \{\sum a[n] \varphi(t - n) : a \in \ell^2\}$, equivalently, there exist constants A, B such that $0 < A \leq \sum_{n \in \mathbb{Z}} |\widehat{\varphi}(w + n)|^2 \leq B < \infty$. The dual function φ_d is defined by:

$$\widehat{\varphi}_d(w) := \frac{\widehat{\varphi}(w)}{\sum_{n \in \mathbb{Z}} |\widehat{\varphi}(w + n)|^2}. \quad (8)$$

Discrete filters are either described by their impulse response $g[n]$ or by their z -transform $G(z) = \sum_{n \in \mathbb{Z}} g[n] z^{-n}$ for which we use upper-case symbols.

Associated with the kernel φ , two natural numbers p and q , and q FIR filters g_0, g_1, \dots, g_{q-1} , we consider the approximation scheme:

$$a[n] = \sum_{j=0}^{q-1} \sum_{k \in \mathbb{Z}} f(qkT - jT) g_j[n - kp], \quad (9)$$

$$f_{\text{approx}}^T(t) = Q_T f(t) := \sum_{n \in \mathbb{Z}} a[n] \varphi\left(\frac{r}{T}t - n\right). \quad (10)$$

where r denotes the ratio $r := p/q$.

Thus, to approximate a function $f(t)$ from its samples $f(nT)$, we first prefilter the samples with the filter bank shown in figure 1, which yields to a coefficient sequence $a[n]$ and then we apply $f_{\text{approx}}^T(t) = \sum a[n]\varphi(rt/T - n)$.

Another possibility, yielding to the same approximation f_{approx}^T , is to evaluate

$$f_{\text{approx}}^T(t) = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} f(qnT - jT) S_j\left(\frac{r}{T}t - np\right),$$

where the reconstruction functions are defined by

$$S_j(t) := \sum_{n \in \mathbb{Z}} g_j[n]\varphi(t - n), \quad j = 0, 1, \dots, q-1. \quad (11)$$

When p and q are coprime, the prefiltering (9) can be also represented as (see figure 2)

$$a[n] = \sum_{k \in \mathbb{Z}} f(kT)h(qn - pk) \quad (12)$$

where $H(z)$ is defined by

$$H(z) := \sum_{j=0}^{q-1} z^{-jp} G_j(z^q). \quad (13)$$

The proof is given in the appendix. There, it is also proved that (13) is equivalent to

$$g_j[n] = h[nq + jp], \quad n \in \mathbb{Z}, j = 0, \dots, q-1. \quad (14)$$

Hence, for any FIR filter $h[n]$, the prefiltering (12) can be represented as (9), where the filters $g_j[n]$ are defined by (14).

See references [17], [18] for efficient structures for fractional decimation and for filter banks.

The (type 2) polyphase components $R_{j,k}$ of the filters G_j are defined by:

$$G_j(z) = \sum_{k=0}^{p-1} z^{-(p-1-k)} R_{k,j}(z^p), \quad j = 0, 1, \dots, q-1. \quad (15)$$

Note that, under the assumed hypotheses, $f_{\text{approx}}^T(t)$ is well defined for any function f . Besides, the reconstruction functions S_j have compact support.

The ratio r plays an important role in the numerical efficiency of the reconstruction algorithm (9)-(10). Indeed, the computational cost of the second step of the algorithm (to evaluate $\sum a[n]\varphi(rt/T - n)$) is proportional to the number of coefficients $a[n]$ considered, which is approximately equal to $r \times N$, where N is the number of available samples (assuming that N is much bigger than the size of the support of the filters g_j). In the next section we show that if we choose the filters with the smallest support, the computational cost of the first step (the prefiltering) of the algorithm is also proportional to r . The cost of this first step is almost negligible compared to the second one (see [1], [5]) which requires evaluations of the kernel.

Note that except in the case of $q = 1$, the system does not satisfy the shift invariant property $Q_T[f(\cdot - Tn)](t) = Q_T f(t - Tn)$ but it satisfies $Q_T[f(\cdot - Tqn)](t) = Q_T f(t - Tqn)$.

III. THE QUASI-INTERPOLATION CONDITION

We denote $\mathbf{G}(z) := [G_0(z), G_1(z), \dots, G_{q-1}(z)]^\top$. As usual, $\xi(w) = O(w^L)$ means that $|\xi(w)/w^L|$ is bounded as $w \rightarrow 0$ (when $\xi \in C^\infty(\mathbb{R})$, it is equivalent to $\xi^{(l)}(0) = 0$ for $l = 0, 1, \dots, L-1$); δ_n denotes the delta sequence; and \bar{c} denotes the conjugate of $c \in \mathbb{C}$. Finally, let $\mathbf{e}_l(z)$, for $l = p, q$, denotes

$$\mathbf{e}_l(z) = [1, z^{-1}, \dots, z^{-(l-1)}]^\top.$$

Recall that we have assumed that the kernel $\varphi(t)$ satisfies the Strang-Fix conditions of order L (5).

The next lemma gives a necessary and sufficient condition on the approximation scheme (9)-(10) in order to reproduce polynomials of degree at most $L-1$.

Lemma 1: The approximation operator Q_T defined in (9)-(10) satisfies $Q_T \rho = \rho$ for all $\rho \in \Pi_{L-1}$ if and only if the condition:

$$\mathbf{e}_q^\top(e^{2\pi i r w}) \mathbf{G}(e^{2\pi i(w + \frac{n}{p})}) \widehat{\varphi}\left(w + \frac{n}{p}\right) - p \delta_n = O(w^L) \quad (16)$$

is satisfied for all $n \in \mathbb{Z}$.

The proofs of this and the next lemma are given in the appendix. Next, we give a more manageable sufficient condition. Notice that, since φ has compact support, the function $A_\varphi(w) = \sum |\varphi(w+n)|^2$ is a trigonometric polynomial (see [19]), and then the function $\widehat{\varphi}_d$ defined in (8) is infinitely differentiable.

Lemma 2: Any of the equivalent conditions:

$$\mathbf{e}_q^\top(e^{2\pi i r w}) \mathbf{G}(e^{2\pi i(w + \frac{k}{p})}) - p \widehat{\varphi}_d(w) \delta_k = O(w^L), \quad (17)$$

for $k = 0, 1, \dots, p-1$, or

$$\left[\frac{d^l}{dw^l} \sum_{j=0}^{q-1} R_{k,j}(e^{2\pi i p w}) e^{-2\pi i(jr + p-1-k)w} \right]_{w=0} = \widehat{\varphi}_d^{(l)}(0) \quad (18)$$

for $k = 0, 1, \dots, p-1$, and for $l = 0, 1, \dots, L-1$, where $R_{k,j}(z)$ are the polyphase components defined in (15), implies that condition (16) holds or equivalently that $Q_T \rho = \rho$ for all $\rho \in \Pi_{L-1}$.

For the case of $r = p = q = 1$, condition (17) reduces to

$$\mathbf{G}(e^{2\pi i w}) = \widehat{\varphi}_d(w) + O(w^L),$$

and it can be found in [7], [11], [12]. Thus, in order to find a FIR filter $g = g_0$, so that the approximation scheme (3) reproduces the polynomials in Π_{L-1} , we have to solve a linear system with L equations whose unknowns are the non null $g[n]$. Thus, in general, it is possible to find a FIR filter g with support of size L satisfying the reproducing condition. If we consider N samples, this filter implies approximately $L \times N$ sums and products in the prefiltering step (assuming that N is much bigger than L).

In the general case, in order to find q FIR filters g_0, g_1, \dots, g_{q-1} , such that the approximation scheme (9)-(10) reproduces the polynomials of Π_{L-1} , we have to solve a linear system with $p \times L$ equations (see (17) or (18)) whose unknowns are the non null $g_j[n]$. Thus, it is possible, in

general, to find FIR filters g_j , being the sum of the size of their supports equal to $p \times L$, and satisfying the reproducing condition. For these filters, the computation of the prefiltering step (9) involves approximately $r \times L \times N$ products and sums; hence the resulting computational cost of the prefiltering step is approximately proportional to r .

If the kernel φ is even (recall that we have supposed that it is real), a suitable property (see [4]), then $\widehat{\varphi}_d^{(l)}(t)$ is real and even, and thus $\widehat{\varphi}_d^{(l)}(0)$ is real and is equals to 0 when l is odd. As a consequence, the equations (18) have real solutions, which implies real filters.

In the case of $r = 1/q$, condition (17) reads as:

$$\left[\frac{d^l}{dw^l} H(e^{2\pi i \frac{w}{q}}) \right]_{w=0} = \widehat{\varphi}_d^{(l)}(0), \quad l = 0, \dots, L-1, \quad (19)$$

where $H(z)$ is given by (13). Note that if the kernel φ is even then any even filter $h(n)$ satisfies the above equations for l odd. Thus, we can achieve symmetric approximation schemes, with a filter of support L when L is odd, and of support $L-1$ when L is even.

As the reader can observe, in the proof of Lemma 2, conditions (17) and (18) are almost always necessary for the reproduction property: $Q_T \rho = \rho$ for all $\rho \in \Pi_{L-1}$.

IV. EXAMPLES WITH THE QUADRATIC B-SPLINE

In order to illustrate the theoretical results we consider as an example the generator

$$\varphi(t) = \beta^2(t) = \begin{cases} \frac{3}{4} - t^2, & -\frac{1}{2} \leq t < \frac{1}{2} \\ \frac{1}{2} (|t| - \frac{3}{2})^2, & \frac{1}{2} \leq |t| < \frac{3}{2} \\ 0, & |t| \geq \frac{3}{2}, \end{cases}$$

the centered quadratic B-spline, which satisfies the Strang-Fix conditions of order $L = 3$. We have (see [7])

$$\widehat{\varphi}_d(w) = \frac{\sin^3(\pi w)}{\pi^3 w^3 [1 - \sin^2(\pi w) + (2/15) \sin^4(\pi w)]}$$

In the case of $r = 1$, the unique filter of the type $H(z) = a + b(z + z^{-1})$, that satisfies the equations (19) is

$$H_1(z) := \frac{5}{4} - \frac{1}{8}(z + z^{-1}). \quad (20)$$

The approximation algorithm, (12) and (10), reads:

$$\begin{aligned} a[n] &= \frac{5}{4} f(nT) - \frac{1}{8} [f(nT + T) + f(nT - T)] \\ f_{\text{approx}}^T(t) &= \sum_{n \in \mathbb{Z}} a[n] \beta^2\left(\frac{t}{T} - n\right). \end{aligned} \quad (21)$$

In the case of $r = 1/2$, the filter

$$H_{1/2}(z) := 2 - \frac{1}{2}(z + z^{-1}) \quad (22)$$

is a solution of (19). The approximation algorithm now reads:

$$\begin{aligned} a[n] &= 2f(2nT) - \frac{1}{2} [f(2nT - T) + f(2nT + T)] \\ f_{\text{approx}}^T(t) &= \sum_{n \in \mathbb{Z}} a[n] \beta^2\left(\frac{t}{2T} - n\right). \end{aligned} \quad (23)$$

In the case of $r = 2/3$, The vector of filter

$$\mathbf{G}_{2/3}(z) := \begin{bmatrix} \frac{25}{16} - \frac{13}{64}(z + z^{-1}) \\ \frac{45}{64}z^{-1} - \frac{9}{32} \\ -\frac{9}{32}z^{-2} + \frac{45}{64}z^{-1} \end{bmatrix} \quad (24)$$

satisfies condition (17). The algorithm (9)-(10) reads:

$$\begin{aligned} a[2n] &= \frac{25}{16} f(3nT) - \frac{9}{32} [f(3nT + T) + f(3nT - T)] \\ a[2n + 1] &= -\frac{13}{64} (f(3nT) + f(3Tn + 3T)) \\ &\quad + \frac{45}{64} [f(3nT + T) + f(3nT + 2T)] \\ f_{\text{approx}}^T(t) &= \sum_{n \in \mathbb{Z}} \left[a[2n] \beta^2\left(\frac{2t}{3T} - 2n\right) \right. \\ &\quad \left. + a[2n + 1] \beta^2\left(\frac{2t}{3T} - 2n - 1\right) \right]. \end{aligned}$$

In order to fix the support of the entries of $\mathbf{G}_{2/3}$ we have taken into account that $a[2n+1]$ is the coefficient of $\beta^2((2/(3T))[t - 3nT - 3T/2])$ which is centered at $3nT + 3T/2$ and $a[2n]$ is the coefficient of $\beta^2((2/(3T))[t - 3nT])$ which is centered at $3nT$.

In the case of $r = 3/4$, the vector of filters

$$\mathbf{G}_{3/4}(z) := \begin{bmatrix} \frac{13}{9} - \frac{17}{81}(z + z^{-1}) \\ -\frac{2}{9} + \frac{26}{27}z - \frac{10}{81}z^2 \\ \frac{10}{27}(z + z^2) \\ -\frac{10}{81}z + \frac{26}{27}z^2 - \frac{2}{9}z^3 \end{bmatrix} \quad (25)$$

satisfies (17). Its corresponding prefiltering step is

$$\begin{aligned} a[3n] &= \frac{13}{9} f(4Tn) - \frac{2}{9} (f(4Tn - T) + f(4nT + T)) \\ a[3n + 1] &= -\frac{17}{81} f(4Tn) - \frac{10}{81} f(4Tn + 3T) \\ &\quad + \frac{10}{27} f(4Tn + 2T) + \frac{26}{27} f(4Tn + T) \\ a[3n + 2] &= -\frac{17}{81} f(4Tn + 4T) + \frac{26}{27} f(4Tn + 3T) \\ &\quad + \frac{10}{27} f(4Tn + 2T) - \frac{10}{81} f(4Tn + T). \end{aligned}$$

Let us recall that the interpolating filter (such that (4) is satisfied) for the generator $\varphi = \beta^2$ is

$$H^{\text{int}}(z) := \frac{1}{\frac{3}{4} + \frac{1}{8}(z + z^{-1})}. \quad (26)$$

Finally, let us give an interesting example from [13] of an IIR filter satisfying the quasi-interpolating condition which can be implemented with a fast algorithm. This example is for the case of $\varphi = \beta^2$ and $r = 1/2$. It is

$$H^{\text{IIR}}(z) := \frac{z^{-1} + z + 2}{\frac{5}{2} + \frac{3}{4}(z^2 + z^{-2})}. \quad (27)$$

In [13] can be found examples for other B-splines β^n and other values of $r = 1/q$.

V. THE APPROXIMATION ERROR

We split this section into three subsections. In the first one we obtain a pointwise estimation for the approximation error and we study the L^∞ -norm of this error. In the second one, we apply the results in [10] to study the L^2 -norm of the error. The third one is concerned with designing the filters G_j in order to get a smaller L^2 -approximation error.

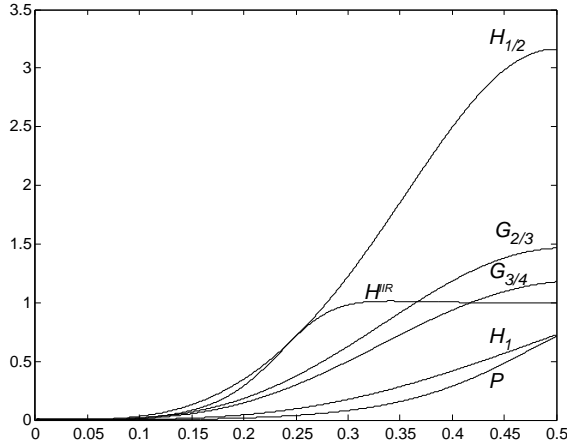


Fig. 3. $\sqrt{E(w/r)}$ for H_1 , $G_{3/4}$, $G_{2/3}$, $H_{1/2}$, H^{IR} and for the orthogonal projector P ($r = 1$ in this case).

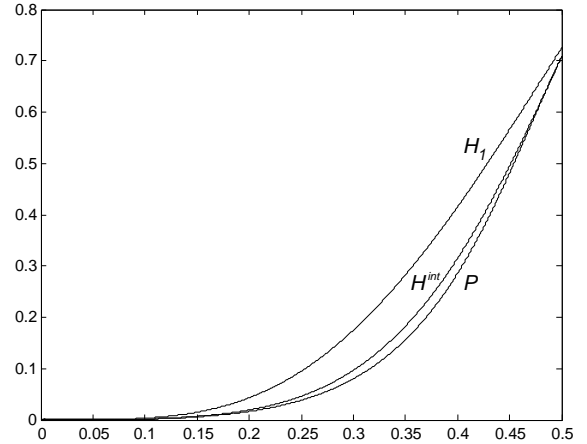


Fig. 4. $\sqrt{E(w/r)}$ for the orthogonal projector P , for the interpolating filter H^{int} and for H_1

A. A pointwise estimation and the L^∞ -approximation error

Theorem 1 proves that, whenever the quasi-interpolation condition (16) holds, the approximation scheme has approximation order L in the L^∞ -norm sense. Besides, it gives a pointwise estimation for the error, which shows that the error depends on the compression factor r . We denote

$$\Omega(t) := (-1)^L \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} (t - pn + jr)^L S_j(t - pn),$$

$$\tilde{\Omega}(t) := \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} |(t - pn + jr)^L S_j(t - pn)|.$$

Note that, since the reconstruction functions S_j are continuous and compactly supported, Ω and $\tilde{\Omega}$ are bounded functions.

Theorem 1: Whenever the condition (16) holds, for all $f \in W_\infty^L(\mathbb{R})$, we have

$$\|f_{\text{approx}}^T - f\|_\infty \leq \frac{\|\tilde{\Omega}\|_\infty \|f^{(L)}\|_\infty}{L!} \frac{T^L}{r^L} \quad (28)$$

and, for all $f \in W_\infty^{L+1}(\mathbb{R})$, we have

$$f_{\text{approx}}^T(t) - f(t) = \frac{\Omega(rt/T) f^{(L)}(t)}{L!} \frac{T^L}{r^L} + O(T^{L+1}). \quad (29)$$

The proof, based on the Taylor polynomial of f , is given in the appendix. For the case of $r = p = q = 1$, the estimations (28) and (29) are given in [8].

B. The L^2 -approximation error

In [10], Blu and Unser gave a formula for the L^2 -norm of the approximation error valid to an ample set of approximation operators \mathcal{L}_h : For any $f \in W_2^k(\mathbb{R})$,

$$\|f - \mathcal{L}_h f\|_2 = \left[\int_{-\infty}^{\infty} |\hat{f}(w)|^2 E(hw) dw \right]^{1/2} + O(h^k), \quad (30)$$

where the integral kernel $E(w)$ depends on the approximation scheme, but does not depend on f . Moreover,

$\left[\int |\hat{f}(w)|^2 E(hw) dw \right]^{1/2}$ gives exactly the average approximation error (see [10, Theorem 2]). This error formula (30), can also be applied to the operator Q_T by changing the sampling step; specifically it can be applied to $\mathcal{L}_h := Q_{hr}$. In this case the formula (30) can be expressed as

$$\|f - Q_T f\|_2 = \left[\frac{1}{T} \int_{-\infty}^{\infty} \left| \hat{f}\left(\frac{w}{T}\right) \right|^2 E\left(\frac{w}{r}\right) dw \right]^{\frac{1}{2}} + O(h^k), \quad (31)$$

where the integral kernel is (assuming that the filters g_j are real):

$$E(w) = 1 - \frac{2}{p} \Re \left[\mathbf{e}_p^\top(z^r) \mathbf{G}(z) \hat{\varphi}(w) \right] + \frac{1}{p^2} \sum_{k=0}^{p-1} \left| \mathbf{e}_q^\top(z^r) \mathbf{G}(ze^{2\pi i \frac{k}{p}}) \right|^2 A_\varphi\left(w + \frac{k}{p}\right) \quad (32)$$

where $z = e^{2\pi i w}$, \Re denotes the real part and $A_\varphi(w) := \sum_{n \in \mathbb{Z}} |\hat{\varphi}(w + n)|^2$ (see the details in the appendix). Figure 3 shows the square root of the integral kernel $\sqrt{E(w/r)}$, corresponding to the orthogonal projector P (see infra (35)), and to the filters H_1 , $G_{3/4}$, $G_{2/3}$, $H_{1/2}$ and H^{IR} defined in (20), (25), (24), (22) and (27) respectively. Figure 4 shows $\sqrt{E(w/r)}$ for P , H_1 , and for the interpolating filter H^{int} defined in (26).

In [10] some results about the approximation error of the operator \mathcal{L}_h are deduced from the formula (30). In Theorem 2 we collect the applications of these results to the operator $Q_T = \mathcal{L}_{T/r}$. In particular, we deduce that the quasi-interpolation condition (16) in Lemma 1 is necessary and sufficient to the approximation scheme (9)-(10) to have order L in the L^2 -norm sense.

Note that since we have assumed that φ has compact support (thus $A_\varphi(w)$ is a trigonometric polynomial) and $G_j(z)$ are FIR filters, the integral kernel $E(w)$ given in (32), has infinite bounded derivatives.

Theorem 2: Assume that g_j are real FIR filters. The quasi-

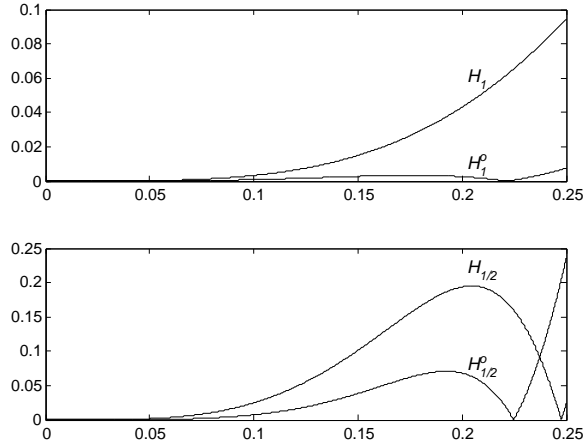


Fig. 5. $\sqrt{E_{\text{res}}(w/r)}$ for H_1 , H_1^o , $H_{1/2}$ and $H_{1/2}^o$.

interpolating condition (16) holds if and only if

$$\|f_{\text{approx}}^T - f\|_2 = O(T^L), \quad \text{for all } f \in W_2^L(\mathbb{R}).$$

Moreover, in this case, for all $f \in W_2^{L+1}(\mathbb{R})$,

$$\|f_{\text{approx}}^T - f\|_2 = \frac{T^L}{rL} \frac{\|f^{(L)}\|_2}{p(2\pi)^L L!} \sqrt{\sum_{n \in \mathbb{Z}} |\Gamma_n^{(L)}(0)|^2} + O(T^{L+1}) \quad (33)$$

where $\Gamma_n(w) = \mathbf{e}_q^\top(e^{2\pi i r w}) \mathbf{G}(e^{2\pi i(w+n/p)}) \widehat{\varphi}(w+n/p)$, and for all $f \in W_2^k(\mathbb{R})$, with $k \geq L$,

$$\|f_{\text{approx}}^T - f\|_2 \leq C \frac{T^L}{rL} \|f^{(L)}\|_2 + K \frac{T^k}{r^k} \|f^{(k)}\|_2 \quad (34)$$

where

$$C = \frac{\sqrt{\|E^{(2L)}\|_\infty}}{2^L \pi^L \sqrt{(2L)!}} \quad \text{and} \quad K = \frac{2p^k}{\pi^k} \left[\|E\|_\infty \sum_{n \geq 1} n^{-2k} \right]^{1/2}.$$

To prove these results we only have to apply the Theorems in [10] (see the appendix). When $r = 1/q$, the above expression for Γ_n reduces to $\Gamma_n(w) = H(e^{2\pi i w/q}) \widehat{\varphi}(w+n)$. In the case of $r = p = q = 1$, it reduces to $\Gamma_n(w) = \widehat{S}(w+n)$. For this case, the estimation (33) is given in [8].

C. Optimal filters

The orthogonal projector onto $V_h(\varphi)$ defined by:

$$P_h f(t) := \frac{1}{h} \sum_{n \in \mathbb{Z}} \int_{-\infty}^{\infty} f(\tau) \varphi_d\left(\frac{\tau}{h} - n\right) d\tau \varphi\left(\frac{t}{h} - n\right) \quad (35)$$

gives the best approximation, in the L^2 -norm sense, by a function in $V_h(\varphi)$. This approximation cannot be computed from the samples of the function. However, we can take advantage of the flexibility provided by the quasi-interpolation schemes, to get an approximation whose error is close to that given by the orthogonal projector. To do this, we will base, as it is done in [7], [12], [13], [14], on the error formula (30).

The orthogonal projector P_h satisfies the estimation (30) and the integral kernel, which we denote by E_{min} (see [7]), is

$$E_{\text{min}}(w) = 1 - \frac{|\widehat{\varphi}(w)|^2}{A_\varphi(w)}.$$

The integral kernel (32) can be expressed as:

$$E(w) = E_{\text{min}}(w) + E_{\text{res}}(w) \quad (36)$$

where

$$E_{\text{res}}(w) = A_\varphi(w) \left| \widehat{\varphi}_d(w) - \frac{1}{p} \mathbf{e}_q^\top(z^r) \mathbf{G}(z) \right|^2 + \frac{1}{p^2} \sum_{k=1}^{p-1} \left| \mathbf{e}_q^\top(z^r) \mathbf{G}(ze^{2\pi i \frac{k}{p}}) \right|^2 A_\varphi\left(w + \frac{k}{p}\right), \quad (37)$$

and $z = e^{2\pi i w}$. The quasi-interpolation condition (17) gives

$$\mathbf{e}_q^\top(z^r) \mathbf{G}(ze^{2\pi i \frac{k}{p}}) - p \widehat{\varphi}_d(w) \delta_k = O(w^K), \quad k = 0, \dots, p-1 \quad (38)$$

holds for $K = L$ and then $E_{\text{res}}(w) = O(|w|^{2L})$. Thus, whenever most of the energy of $f(w/T)$ is concentrated close to zero, the approximation error is small. In order to do it smaller we can try to find a vector of filters satisfying (38) for a bigger K . Besides, having in mind that a part of the energy of $f(w/T)$ could not be so close to zero, we can also use the flexibility given for a bigger support of the filters in order to minimize (see (31) and (36))

$$\int_{-\infty}^{\infty} v(w) E_{\text{res}}\left(\frac{w}{r}\right) dw \quad (39)$$

where $v(w) \geq 0$ is a suitable weight. Namely, we fix a weight $v(w)$, an order $K \geq L$ and the support of the filters. Then, we search the \mathbf{G} that minimize (39) among those which satisfy (38). This is a linear problem easy to solve. For example, when $\varphi = \beta^2$, $r = 1$ the symmetrical filters $H(z)$ of the type

$$H(z) = a + \frac{b}{2}(z + z^{-1}) + \frac{c}{2}(z^2 + z^{-2})$$

satisfying (38) for $K = L + 1 = 4$ are those given by

$$a = \frac{5}{4} + 3c, \quad b = -\frac{1}{4} - 4c.$$

Thus, minimizing (39) is equivalent to finding c such that

$$\int_{-\infty}^{\infty} v(w) A_\varphi(w) \left(\widehat{\varphi}_d(w) - \frac{5}{4} + \frac{1}{4} \cos(2\pi w) - c [3 - 4 \cos(2\pi w) + \cos(4\pi w)] \right) dw$$

is minimum. For instance, when $v(w)$ is the characteristic function of the interval $(-1/4, 1/4)$ then $c \approx 0.054$ and the searched filter is

$$H_1^o(z) := 1.412 - 0.233(z + z^{-1}) + 0.027(z^2 + z^{-2}).$$

For the rate $r = 1/2$ the same criterium provides

$$H_{1/2}^o(z) := 2.87 - 1.08(z + z^{-1}) + 0.145(z^2 + z^{-2}).$$

Figure 5 shows the square root of $E_{\text{res}}(w/r)$ for these filters as well as for those with minimum support. Note that the bigger support increases the computations in the prefiltering step, but has no influence on the number of evaluations of the kernel.

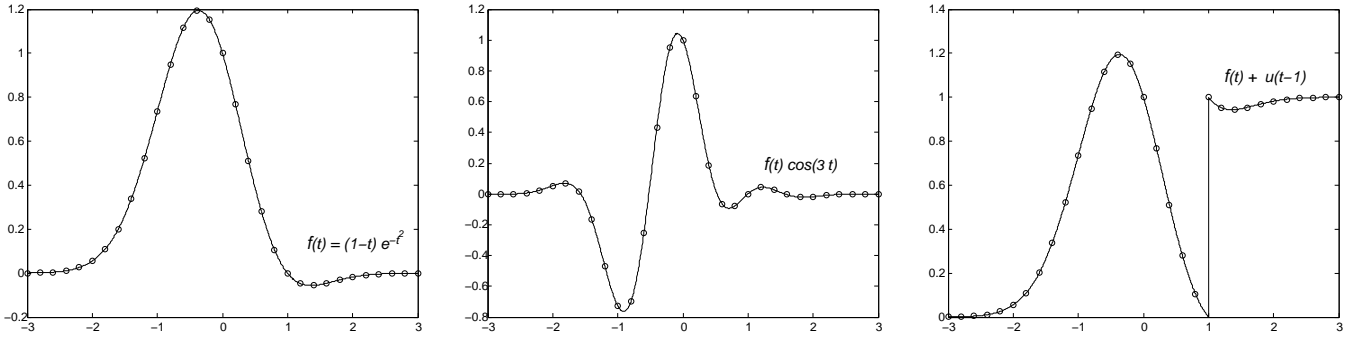


Fig. 6. Examples

	H^{int}	H_1^o	H_1	$G_{3/4}$	$G_{2/3}$	H^{IIR}	$H_{1/2}^o$	$H_{1/2}$
$f(t)$	3.8×10^{-4}	4×10^{-4}	7×10^{-4}	2.6×10^{-3}	3.2×10^{-3}	3.6×10^{-3}	4×10^{-3}	6.4×10^{-3}
$f(t)\cos(3t)$	3.2×10^{-3}	3.1×10^{-3}	7.2×10^{-3}	2.3×10^{-2}	3.2×10^{-2}	4.1×10^{-2}	4.4×10^{-2}	6.1×10^{-2}
$f(t) + u(t-1)$	0.26	0.26	0.25	0.2	0.25	0.21	0.47	0.26

 TABLE I
 $L^2[-3, 3]$ -NORM OF THE ERROR

VI. SIMULATIONS

In this section we give numerical simulations showing the behavior of the approximation scheme given in Sections IV and V. Specifically, we apply the algorithms to recover the functions shown in figure 6 from 31 samples taken in the interval $[-3, 3]$ with period $T = 0.2$. The $L^2[-3, 3]$ -norm of the error for the different algorithms and examples are given in Table I.

For instance for the first example $f(t) = (1-t)e^{-t^2}$, the algorithm (21) that does not entail compression H_1 (corresponding to $r = 1$), gives an error of 7×10^{-4} . It needs 30 evaluations of the kernel. The algorithm corresponding to $H_{1/2}$ needs only 15 evaluations of the kernel, but its error increases to 6.4×10^{-3} . Between these cases, are those corresponding to the rates $r = 3/4$ and $r = 2/3$.

For the second example the errors are bigger (it has wide variations). The behavior with respect to the different rates are the same.

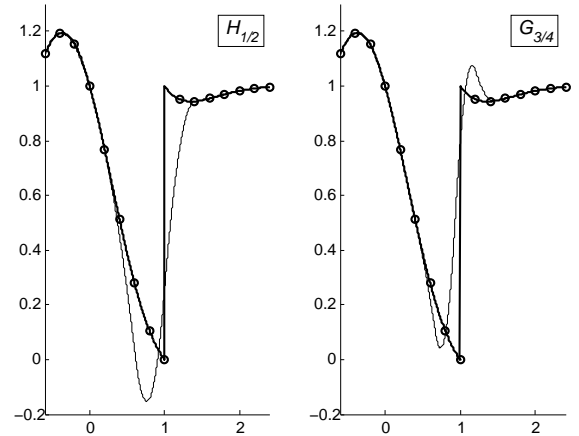
For the third example, the $L^2[-3, 3]$ -norm of the error, due fundamentally to the discontinuity, is approximately equal for all the rates. However the reproduction of the discontinuity is better using a bigger r as it is shown in figure 7.

Table I also gives the errors for the interpolating filter H^{int} and for the IIR filter H^{IIR} , which are close to those given by H_1 and $H_{1/2}^o$ respectively.

The pointwise error for the second example and for some of the algorithms can be seen in figure 8.

As can be observed in Table I, using the optimized filters H_1^o and $H_{1/2}^o$ improves the results with respect to the minimum support filters H_1 and $H_{1/2}$. The reconstruction of the second example using H_1 and H_1^o and sampling period $T = 0.5$ can be seen in figure 9.

The gradient, used in some applications (e.g. contour detection, shading), can be computed by means of the coefficients


 Fig. 7. Function $f(t) + u(t-1)$ and the reconstructed functions from samples taken with period $T = 0.2$ using $H_{1/2}$ and $G_{3/4}$ respectively

$a[n]$ obtained in the prefiltering step (9). Indeed, using (10) and that $\frac{d}{dt}\beta^2(t) = \beta^1(x+1/2) - \beta^1(x-1/2)$, where β^1 is the B-spline of degree 1, we obtain that

$$\frac{d}{dt}f_{\text{approx}}^T(t) = \sum_{n \in \mathbb{Z}} \frac{r}{T} (a[n] - a[n-1]) \beta^1\left(\frac{rt}{T} - n + \frac{1}{2}\right).$$

For the first example, $T = 0.2$, and using H_1 , $G_{3/4}$, $G_{2/3}$ and $H_{1/2}$ this formula gives $L^2[-3, 3]$ -norm error equal to 1.1×10^{-2} , 2.1×10^{-2} , 2.7×10^{-2} and 5.1×10^{-2} respectively.

VII. CONCLUSION

In this work we have studied the approximation of functions from their samples by using quasi-interpolation formulas. These schemes give a good quality since they have a certain

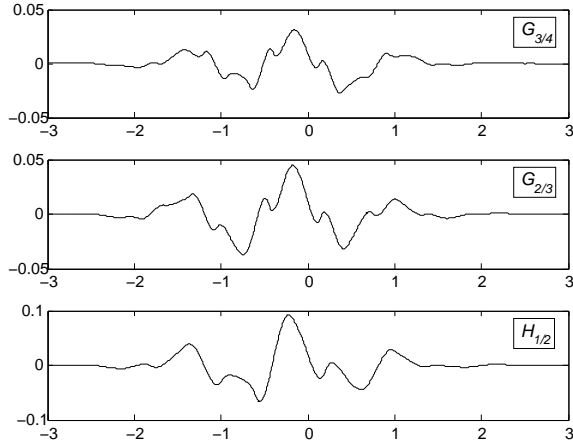


Fig. 8. Pointwise error in the reconstruction of the second example $f(t) \cos(3t)$ using $G_{3/4}$, $G_{2/3}$ and $H_{1/2}$

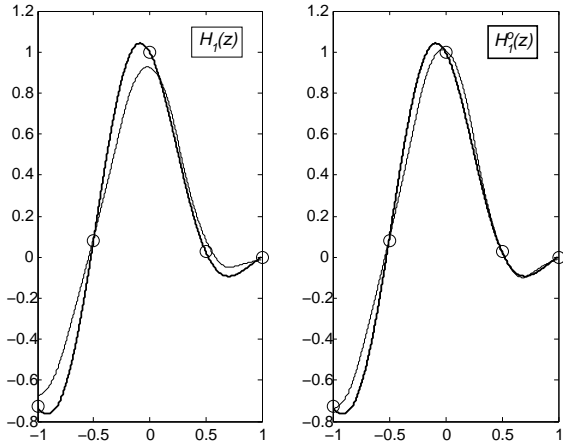


Fig. 9. Function $f(t) \cos(3t)$ and the reconstructed functions from samples taken with period $t = 0.5$ using the filters H_1 and H_1^o respectively

order of approximation and they are numerically efficient since the necessary prefiltering of the samples can be done with little support. In particular, they give a quality close to that given by the classical interpolating formula.

A rational filter bank with a rate $r < 1$ can be used for the prefiltering of the samples, maintaining the approximation order and the little support. When the rate r decrease, we lose in quality of approximation but we gain in numerical efficiency. Thus, in applications where the speed is more relevant than the quality, it is advisable to use a rate $r < 1$.

APPENDIX

In this appendix we include the proofs of the results in Sections II, III and V.

Proof of that (9) can be represented by (12) and of that (13) and (14) are equivalent: First, we prove that for any $k \in \mathbb{Z}$, there exist unique numbers n and j satisfying

$$k = nq + jp, \quad n \in \mathbb{Z}, \quad j \in \{0, 1, \dots, q-1\}. \quad (40)$$

Indeed, since q and p are coprime, there exist $a, b \in \mathbb{Z}$ such that $1 = ap + bq$. Then $k = kap + kbq$. Using that ka can be written as $ka = lq + j$, with $j \in \{0, \dots, q-1\}$ and $l \in \mathbb{Z}$, we obtain (40) with $n = lp + kb$. The numbers n and q are unique since $nq + jp = \tilde{n}q + \tilde{j}p$ implies $(n - \tilde{n})q = (\tilde{j} - j)p$, $\tilde{j} - j \in \{-(q-1), \dots, q-1\}$, and q, p are coprime.

From the representation (40) we obtain that $H(z) = \sum_{j=0}^{q-1} z^{-jp} G_j(z^q) = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} g_j[n] z^{-(nq+jp)}$ if and only if $g_j[n] = h[nq + jp]$, for all $n \in \mathbb{Z}$ and $j \in \{0, \dots, q-1\}$. Hence, (13) and (14) are equivalent.

Using that $g_j[n] = h[nq + jp]$ we obtain

$$\begin{aligned} & \sum_{j=0}^{q-1} \sum_{k \in \mathbb{Z}} f(qkT - jT) g_j[n - kp] \\ &= \sum_{j=0}^{q-1} \sum_{k \in \mathbb{Z}} f([qk - j]T) h[qn - p(qk - j)] \\ &= \sum_{k \in \mathbb{Z}} f(kT) h(qn - pk). \quad \square \end{aligned}$$

Proof of Lemma 1: Notice first that Q_T reproduces the polynomials in Π_{L-1} in and only if Q_r does it, since $Q_T = \sigma_{T/r} Q_r \sigma_{r/T}$ where σ_a denotes the operator $\sigma_a f = f(\cdot/a)$.

Notice that $\sum_{n \in \mathbb{Z}} (pn - jr)^l S_j(t - pn)$ converges in the sense of tempered distributions (see [21, Theorem 4.3-1]). Applying the Fourier transform we obtain that

$$t^l = [Q_r x^l](t) = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} (pn - jr)^l S_j(t - pn)$$

if and only if

$$\frac{\delta^{(l)}(w)}{(-2\pi i)^l} = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} (pn - jr)^l e^{-2\pi i p n w} \hat{S}_j(w)$$

where δ denotes the Dirac delta. The Fourier transform of $(x - jr)^l$ has compact support. Thus, applying the Poisson formula (see [22]) we obtain that the above equality holds if and only if

$$p \delta^{(l)}(w) = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} e^{-2\pi i j (r w - \frac{n}{q})} \delta^{(l)}\left(w - \frac{n}{p}\right) \hat{S}_j(w)$$

or equivalently if

$$\begin{aligned} \delta_n p \delta^{(l)}(w) &= \sum_{j=0}^{q-1} e^{-2\pi i j (r w - \frac{n}{q})} \delta^{(l)}\left(w - \frac{n}{p}\right) \hat{S}_j(w) \\ &= \sum_{j=0}^{q-1} e^{-2\pi i j (r w - \frac{n}{q})} \delta^{(l)}\left(w - \frac{n}{p}\right) G_j(e^{2\pi i w}) \hat{\varphi}(w) \end{aligned}$$

for all $n \in \mathbb{Z}$. Taking into account that for any function $\xi \in C^\infty(\mathbb{R})$ we have $\xi(w) \delta^{(l)}(w - n/p) = 0$, for $l = 0, 1, \dots, L-1$, if and only if $\xi(w + n/p) = O(w^L)$, we deduce that the above equality holds for every $l = 0, 1, \dots, L-1$, or equivalently Q_r reproduces polynomials of Π_{L-1} , if and only if condition (16) is satisfied for all $n \in \mathbb{Z}$. \square

Proof of Lemma 2: Let z denotes $z = e^{2\pi iw}$. A sufficient condition, (also necessary in most of the times) for (16) is

$$\begin{aligned} \mathbf{e}_q^\top(z^r)\mathbf{G}(ze^{2\pi i\frac{n}{p}}) &= O(w^L), \quad n = 1, \dots, p-1, \\ \mathbf{e}_q^\top(z^r)\mathbf{G}(z)\widehat{\varphi}(w) - p &= O(w^L). \end{aligned} \quad (41)$$

Since we have assumed that φ satisfies the Strang-Fix conditions of order L and has compact support, we have

$$1 - \overline{\widehat{\varphi}}_d(w)\widehat{\varphi}(w) = 1 - \frac{|\widehat{\varphi}(w)|^2}{\sum_{n \in \mathbb{Z}} |\widehat{\varphi}(w+n)|^2} = O(w^{2L})$$

(see inequality (53) in [8]). Hence, condition (41) is equivalent to (17).

In terms of the polyphase components (15) the condition (17) reads as, for $n = 0, 1, \dots, p-1$,

$$\sum_{k=0}^{p-1} C_k(w) a^{n(p-1-k)} = p \overline{\widehat{\varphi}}_d(w) \delta_n + O(w^L)$$

where $C_k(w) = \sum_{j=0}^{q-1} R_{k,j}(z^p) z^{-jr-(p-1-k)}$, $z = e^{2\pi iw}$ and $a = e^{-2\pi i/p}$, or equivalently, for $n = 0, 1, \dots, p-1$,

$$\sum_{k=0}^{p-1} C_k^{(l)}(0) a^{n(p-1-k)} = p \overline{\widehat{\varphi}}_d^{(l)}(0) \delta_n, \quad l = 0, \dots, L-1.$$

Denoting $\mathbb{C} = [C_k^{(l)}(0)]_{l=0, \dots, L-1, k=0, \dots, p-1}$ and $\mathbb{A} = [a^{n(p-1-k)}]_{k=0, \dots, p-1, n=0, \dots, p-1}$, the above equalities read as

$$\mathbb{C}\mathbb{A} = p [\overline{\widehat{\varphi}}_d^{(l)}(0) \delta_n]_{l=0, \dots, L-1, n=0, \dots, p-1}$$

Using that $\mathbb{A}\mathbb{A}^* = p\mathbb{I}$, where \mathbb{A}^* denotes the transpose conjugate of \mathbb{A} and \mathbb{I} the identity matrix, we deduce that the above equality holds if and only if $C_k^{(l)}(0) = \overline{\widehat{\varphi}}_d^{(l)}(0)$ for $k = 0, \dots, p-1$, $l = 0, \dots, L-1$, which proves (18). \square

Proof of Theorem 1: Let $P_t^k(x)$ denotes the Taylor polynomial of degree k for f at t . Any function f in the Sobolev space $W_\infty^{L+1}(\mathbb{R})$ satisfies (see [20, Chapter 2])

$$f(x) = P_t^L(x) + R_t^L(x) = \frac{f^{(L)}(t)}{L!} (x-t)^L + P_t^{L-1}(x) + R_t^L(x)$$

where the remainder R_t^L can be expressed as

$$R_t^L(x) = \frac{(x-t)^{L+1}}{L!} \int_0^1 (1-\tau)^L f^{(L+1)}(t+\tau[x-t]) d\tau.$$

Since Q_T reproduces polynomials in Π_{L-1} we have $f(t) = P_t^{L-1}(t) = Q_T P_t^{L-1}(t)$. Hence, using the notation $[Q_T \xi(x)](t)$ for $Q_T \xi(t)$, we obtain

$$\begin{aligned} Q_T f(t) - f(t) &= Q_T (f - P_t^{L-1})(t) \\ &= \left[Q_T \left\{ \frac{f^{(L)}(t)}{L!} (x-t)^L + R_t^L(x) \right\} \right](t) \\ &= \frac{f^{(L)}(t)}{L!} [Q_T (x-t)^L](t) + Q_T R_t^L(t). \end{aligned} \quad (42)$$

We have

$$\begin{aligned} [Q_T (x-t)^L](t) &= \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} (qnT - jT - t)^L S_j \left(\frac{rt}{T} - np \right) \\ &= \frac{T^L}{r^L} \Omega \left(\frac{rt}{T} \right). \end{aligned} \quad (43)$$

Besides, taking into account that

$$|R_t^L(qnT - jT)| \leq \frac{T^{L+1}}{r^{L+1}} \frac{\|f^{(L+1)}\|_\infty}{(L+1)!} \left(\frac{r}{T} t - pn + jr \right)^{L+1},$$

and that $\sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} |t - pn + jr|^{L+1} |S_j(t - pn)|$ is a bounded function we obtain

$$Q_T R_t^L(t) = \sum_{j=0}^{q-1} \sum_{n \in \mathbb{Z}} R_t^L(qnT - jT) S_j \left(\frac{rt}{T} - pn \right) = O(T^{L+1})$$

From this estimation, (42) and (43) the estimation (29) follows.

Similarly we can prove that for $f \in W_\infty^L(\mathbb{R})$, the remainder $R_t^{L-1}(x) = f(x) - P_t^{L-1}(x)$ satisfies

$$|Q_T R_t^{L-1}(t)| \leq \frac{\|\widetilde{\Omega}\|_\infty \|f^{(L)}\|_\infty}{L!} \frac{T^L}{r^L}$$

which, as $Q_T f(t) - f(t) = Q_T (f - P_t^{L-1})(t) = Q_T R_t^{L-1}(t)$, proves (28). \square

Proof that (31) holds with $E(w)$ given by (32): It can be easily verified that the approximation operator Q_T can be expressed by

$$Q_T f(t) = \sum_{n \in \mathbb{Z}} \int f(\tau) \widetilde{\Phi}^\top \left(\frac{\tau}{h} - pn \right) \Phi \left(\frac{t}{h} - np \right) \frac{d\tau}{h} \quad (44)$$

where $h = T/r$, $\Phi = [\phi_0, \dots, \phi_{p-1}]^\top$, $\widetilde{\Phi} = [\widetilde{\phi}_0, \dots, \widetilde{\phi}_{p-1}]^\top$ and

$$\phi_s(t) = \varphi(t-s), \quad \widetilde{\phi}_s(t) = \sum_{j=0}^{q-1} \sum_{k \in \mathbb{Z}} g_j [kp+s] \delta(t+pk+jr)$$

for $s = 0, 1, \dots, p-1$. By applying the Theorem 1 in [10] it follows that $\mathcal{L}_h := Q_{hr}$ satisfies (30), where the kernel is given by

$$\begin{aligned} E(w) &= \frac{1}{p^2} |p - \overline{\widehat{\Phi}}(w)\widehat{\Phi}(w)|^2 + \frac{1}{p^2} \sum_{n \neq 0} \left| \overline{\widehat{\Phi}}(w)\widehat{\Phi}\left(w + \frac{n}{p}\right) \right|^2 \\ &= 1 - \frac{2}{p} \Re[\overline{\widehat{\Phi}}(w)\widehat{\Phi}(w)] + \frac{1}{p^2} \sum_{n \in \mathbb{Z}} \left| \overline{\widehat{\Phi}}(w)\widehat{\Phi}\left(w + \frac{n}{p}\right) \right|^2. \end{aligned} \quad (45)$$

Denoting $z = e^{2\pi iw}$, for $n \in \mathbb{Z}$, we have

$$\begin{aligned} \overline{\widehat{\Phi}}(w)\widehat{\Phi}\left(w + \frac{n}{p}\right) &= \sum_{s=0}^{p-1} \sum_{j=0}^{q-1} z^{-jr} \sum_{k \in \mathbb{Z}} g_j [kp+s] z^{-pk} e^{-2\pi is(w+\frac{n}{p})} \widehat{\varphi}\left(w + \frac{n}{p}\right) \\ &= \sum_{j=0}^{q-1} z^{-jr} \sum_{s=0}^{p-1} \sum_{k \in \mathbb{Z}} g_j [kp+s] (ze^{2\pi i\frac{n}{p}})^{-pk-s} \widehat{\varphi}\left(w + \frac{n}{p}\right) \\ &= \sum_{j=0}^{q-1} z^{-jr} G_j(z e^{2\pi i\frac{n}{p}}) \widehat{\varphi}\left(w + \frac{n}{p}\right) \\ &= \mathbf{e}_q^\top(z^r)\mathbf{G}(ze^{2\pi i\frac{n}{p}}) \widehat{\varphi}\left(w + \frac{n}{p}\right). \end{aligned} \quad (46)$$

Then

$$\begin{aligned} & \sum_{n \in \mathbb{Z}} \left| \widehat{\Phi}(w) \widehat{\Phi}\left(w + \frac{n}{p}\right) \right|^2 \\ &= \sum_{n \in \mathbb{Z}} \left| \mathbf{e}_q^\top(z^r) \mathbf{G}(ze^{2\pi i \frac{n}{p}}) \widehat{\varphi}\left(w + \frac{n}{p}\right) \right|^2 \\ &= \sum_{s=0}^{p-1} \sum_{n \in \mathbb{Z}} \left| \mathbf{e}_q^\top(z^r) \mathbf{G}(ze^{2\pi i \frac{s}{p}}) \widehat{\varphi}\left(w + n + \frac{s}{p}\right) \right|^2 \\ &= \sum_{s=0}^{p-1} |\mathbf{e}_q(z^r) \mathbf{G}(ze^{2\pi i \frac{s}{p}})|^2 A_\varphi\left(w + \frac{s}{p}\right). \end{aligned}$$

Using this equality and (45), the expression for the integral kernel (32) follows. From (30), with $h = T/r$, (31) follows. Notice that by applying the Lemma 2 in [10] using the expression (46), we obtain a new proof of Lemma 1. \square

Proof of Theorem 2: We use the notation of the above proof. As $\sum_{n \in \mathbb{Z}} \sum_{s=0}^{p-1} |x - s - np|^{2L} \phi_s(x - np)$ is bounded (since φ is continuous and has compact support), and $\int |\widetilde{\phi}_s(t)| dt$ and $\int |t - n|^{2L} |\widetilde{\phi}_s(t)| dt$ are bounded (since g_j are FIR filters), the hypothesis of Theorem 3 in [10] holds for $\mathcal{L}_h := Q_{hr}$. By applying this theorem and Lemma 1 the first assertion of Theorem 2 follows. As the integral kernel $E(w)$ has, in this case, infinite bounded derivatives, the estimation (33) follows from Theorem 4 in [10] taking into account (46). Finally the inequality (34) follows from Theorem 5 in [10]. \square

ACKNOWLEDGMENT

The author wish to thank the referees for their valuable and constructive comments and to thank A.G. García for its kind reading of the manuscript and for the suggested changes.

REFERENCES

[1] P. Thévenaz, T. Blu and M. Unser, "Interpolation revisited", *IEEE Trans. Med. Ima.*, vol. 19, no. 7, pp. 739-758, Jul. 2000.
 [2] M. Unser, "Sampling-50 years after Shannon", *Proc. IEEE*, vol. 88, no. 4, pp. 569-587, Apr. 2000.
 [3] A. Aldroubi and M. Unser, "Sampling procedures in funtion spaces and asymptotic equivalence with Shannon's sampling theory", *Numer. Funct. Anal. Optimiz.*, vol. 15, no. 1, pp. 1-21, Feb. 1994.
 [4] P. Thévenaz, T. Blu and M. Unser, "Image interpolation and resampling", in *Handbook of Medical Imaginig Processing and Analysis, I. N. Bankman, Ed. New York Academic 2000*, pp. 393-420.
 [5] T. Blu, P. Thévenaz and M. Unser, "MOMS: Maximal-order interpolation of minimal support", *IEEE Trans. Signal Process.*, vol. 10, no. 7, pp. 1069-1080, Jul. 2001.
 [6] M. Unser, A. Aldroubi and M. Eden, "B-spline signal processing: part II- efficient design and applications.", *IEEE Trans. Signal Process.*, vol. 41, no. 2, pp. 834-848, Feb. 1993.
 [7] T. Blu and M. Unser, "Quantitative Fourier analysis of approximation techniques: Part I-Interpolators and Projectors, and Part II-Wavelets", *IEEE Trans. Signal Process.*, vol. 47, no. 10, pp. 2783-2806, Oct. 1999.
 [8] M. Unser and I. Daubechies, "On the approximation power of convolution-based least squares versus interpolation", *IEEE Trans. Signal Process.*, vol. 45, no. 7, pp. 1697-1711, July 1997.
 [9] J. J. Lei, R. Q. Jia and E. W. Cheney, "Approximation from shift-invariant spaces by integral operators", *SIAM J. Math. Anal.*, vol. 28, no. 2, pp. 481-498, Mar. 1997.
 [10] T. Blu and M. Unser, "Approximation error for quasi-interpolators and (multi-)wavelet expansions", *Appl. Comput. Harmon. Anal.*, vol. 6, no. 2, pp. 219-251, Mar. 1999.
 [11] M. Unser, "Quasi-Orthogonality and Quasi-projections", *Appl. Comput. Harmon. Anal.*, vol. 3, no. 3, pp. 201-251, July 1996.

[12] L. Condat, T. Blu and M. Unser, "Beyond interpolation: Optimal reconstruction by quasi-interpolation", *Proc. IEEE Int. Conf. Image Processing.*, vol. 1, pp. 33-36, Sept. 2005.
 [13] L. Condat, A. Montanvert, "Analyse multirésolution L^2 -optimale: estimation par quasi-projections" *Proc. of GRETSI*, Sept. 2005.
 [14] L. Condat and D. Van De Ville, "Quasi-interpolating spline models for hexagonally-sampled data", *IEEE Trans. Signal Process.*, vol. 16, no. 5, pp. 1195-1206, May 2007.
 [15] A. G. García, M. A. Hernández-Medina and G. Pérez-Villalón, "Over-sampling in shift-invariant spaces with a rational sampling period", *IEEE Trans. Signal Process.*, vol. 57, no. 9, pp. 3442-3449, Sept. 2009.
 [16] L. Condat, A. Montanvert, "Reconstruction from non-uniform samples: A direct variational approach in shift-invariant spaces", *Tech. report hal-00338881*, 2005.
 [17] P. P. Vaidyanathan, *Multirate Systems and Filter Banks*. Prentice Hall, 1993.
 [18] M. Vetterli and J. Kovačević, *Wavelets and Subband Coding*. Prentice Hall, 1995.
 [19] C. K. Chui, *An Introduction to Wavelets*, Academic Press, 1992.
 [20] R. DeVore and G. G. Lorentz *Constructive Approximation*. Springer, 1993.
 [21] A. H. Zemanian, *Distribution Theory and Transform Analysis* Dover, New York, 1965.
 [22] C. Gasquet and P. Witomski, *Fourier Analysis and Applications*. Springer, 1999.



G. Pérez-Villalón was born in Teruel, Spain, in 1960. He received the M.Sc. degree in mathematics in 1983 from the University Complutense of Madrid, Spain, and the Ph.D. in mathematics from the University Carlos III de Madrid, Spain, in 2005. He is currently Associate Professor in the Department of Mathematics of the EUITT in the Madrid Polytechnic University, Spain. His research interests include wavelets, interpolation, approximation, and sampling theory.